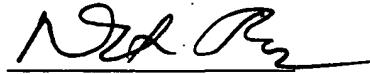




I hereby certify that this correspondence (including Exhibits) is being deposited with the United States Postal Service via Express Mail in an envelope addressed to the Commissioner for Patents, P.O. Box 1450, Alexandria, VA 22313-1450, on August 15, 2006 (Express Mail Label No.: ET615079096US).


Natu J. Patel

IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In re Application of C. Earl Woolfork

Serial No. 10/648,012 : Group Art Unit: 2615

Confirm. No.: 3337 : Examiner: Andrew C. Flanders

Filed: August 26, 2003

For: WIRELESS DIGITAL AUDIO MUSIC SYSTEM

DECLARATION OF APPLICANT REGARDING LIMITED BATTERY LIFE
UNDER 35 USC Section 132

I, C. Earl Woolfork, being duly sworn, depose and declare as follows:

1. I am the Inventor of the above referenced patent application ("Application"). I have personal knowledge of the following matter and if asked to testify, could and would testify competently, thereto.

2. Daphne Burton, my then attorney, conducted the interview with Examiner Flanders and Supervisory Patent Examiner Tran (collectively "Examiners") on June 13, 2006 regarding the pending office action dated May 17, 2006. I participated in that interview.

3. During the interview, among other things, we discussed U.S. Patent No. 5,771,441 issued to Altstatt ("Altstatt" or "the 441 Patent") and U.S. Patent No. 5,946,343 issued to Schotz ("Schotz" or "the 343 Patent").

4. Examiners requested that I submit evidence in an affidavit under 35 USC Section 132 explaining as to why the combination of Altstatt in view of Schotz is non-operative due to limited battery life.

5. I am hereby submitting this affidavit and all the supporting documentation to the Examiners for their consideration.

BEST AVAILABLE COPY

6. Altstatt's invention is based on an analog technology and is operated by a battery. Altstatt recites that the maximum value of V is fixed by the battery voltage of 1.5 or possibly 3 volts (Column 8, lines 22-24).

7. Schotz' invention is based on digital technology. Schotz's digital wireless speaker system requires 120VAC at 60Hz. Schotz further states that "[b]oth the transmitter 22 and the receiver 24 have respective power circuits (not shown) that convert input power (e.g., 120VAC at 60 Hz) into proper voltage levels for appropriate transmitter and receiver operation." Please refer to Column 14, lines 1-4.

8. Exhibit A, attached hereto, lists the commercially available Integrated Chip components ("IC Components") that both Altstatt and Schotz identify in their respective designs. Datasheets identifying electrical current requirements to operate the IC Components are included in Exhibit B.

9. Altstatt cannot be combined with Schotz. However, even assuming such a combination is possible, the Altstatt's battery powered analog headphone system will suffer from a significantly reduced playtime due to the power consumption of Schotz's numerous integrated circuit components, as articulated in the calculation spreadsheet attached hereto as Exhibit C.

10. The "playtime" is defined as the time the invention can be operated continuously before the battery must be changed or recharged. The playtime calculation consists of simple unit conversions as defined in chapter one, problem 1.5 and solution set of well known Theodore S. Rappaport's Wireless Communications Principles & Practice textbook. The relevant pages from the textbook are attached herewith as Exhibit D.

According to Exhibit D, the formula for the playtime calculation is:

$$\{((60\text{minutes}/1\text{hour}) \times B\text{mA-h}) / [(60 \text{ minutes}/\text{hour} \times 24 \text{ hour}/\text{day})(\text{sum of IC currents in mA})]\} \times (24\text{hour}/\text{day})$$

where B is the battery current capacity.

11. As shown in Exhibit C, Altstatt's portable invention will yield a playtime greater than 10 hours when operated with a small battery having a current capacity of 50mA-h (50 milliamp-hours).

12. If we were to hypothetically apply the same 50mA-h battery capacity to operate Schotz's invention, Exhibit C further shows that the frequency hopping spread spectrum ("FHSS") system will operate for approximately six minutes, and the direct sequence spread spectrum ("DSSS") system will operate for approximately eleven

Docket No.: W003-4000

PATENT

minutes before requiring a new battery or a recharged battery. Please note that the FHSS and DSSS system operations are constrained to the lowest device (transmitter or receiver) operation time.

Date: 8/14/06

Respectfully Submitted,

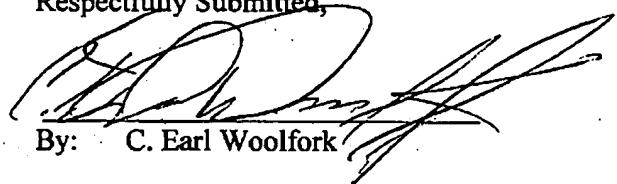

By: C. Earl Woolfork

EXHIBIT A

US Patent Number:5,771,441 Issued to Altstatt

Number	Component Description	Reference
1	Transmitter,BA1404	column 5, lines 34-37
2	Receiver,TA7766AF	column 8, lines 54-58
3	Receiver,TA7792F	column 8, lines 54-58

US Patent Number:5,946,343 Issued to Schotz

1	Digital Signal Processor,DSP56002	column 14, lines 49-50
2	A/D converter,SAA7360	column 7, lines 11-12
3	Stereo Filter MPEG,SAA2520	column 14, lines 47-48
4	MPEG,SAA2521	column 14, lines 47-48
5	Modulator,RF2422	column 10, lines 17-18
6	Power Amplifier,TQ9132	column 10, lines 31-32
7	Phase Locked Loop,MC12210	column 10, lines 49-50
8	Voltage Controlled Oscillator,SMV2500	column 14, lines 51-53
9	Low Noise Amplifier,MGA86576	column 11, lines 16-18
10	Digital Interface Transmitter,CS8402	column 11, lines 31-33
11	Digital to Analog Converter,TDA1305T	column 13, lines 57-59
12	Clock Recovery & Timing,TRU-050	column 12, lines 28-29
13	Demodulator,RF2703	column 12, lines 13-15
14	Microprocessor,PIC16C55	column 6, lines 63-66
15	DSSS Transmitter,CYLINK SSTX	column 16, lines 62-64
16	DSSS Receiver,CYLINK Part#SPECTRE	column 18, lines 4-5
17	Mixer,IAM81008	column 11, lines 16-18
18	Channel Encoder/Decoder,SRT241203	column 9, lines 25-26
19	Interleaver/De-interleaver,SRT-24INT	column 9, lines 50-52
20	Optical Digital Receiver,HK-3131-01	column 7, lines 40-43
21	Optical Digital Transmitter,HK-3131-03	column 13, lines 15-17
22	Voltage Controlled Oscillator,M2 D300	column 8, lines 49-50

EXHIBIT B

US Patent Number:5,771,441 Issued to Altstatt

Item Number 1: Transmitter, BA1404

ROHM CO LTD

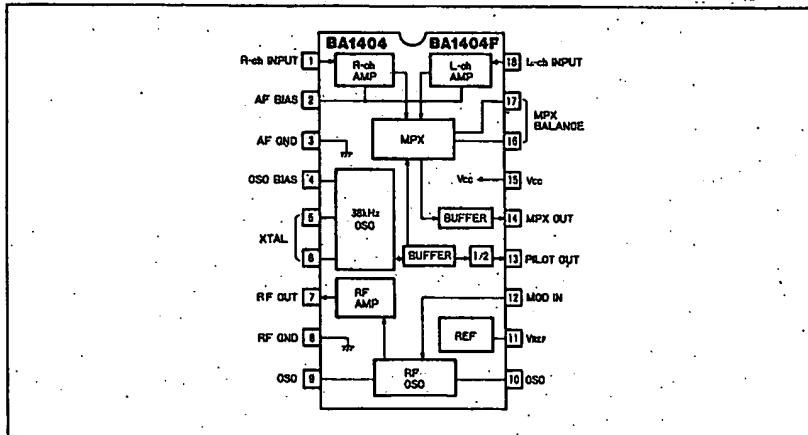
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オーディオ用 IC / ICs for Audio Applications

BA1404/BA1404F

● ブロックダイアグラム/Block Diagram

T-77-05-05



● 绝对最大定格 / Absolute Maximum Ratings ($T_a=25^\circ\text{C}$)

Parameter	Symbol	Limits	Unit
電源電壓	V _{CO}	2.6	V
許容損失	P _d	600*	mW
動作溫度範圍	T _{opr}	-25~75	°C
保存溫度範圍	T _{stg}	-60~125	°C

* $T_B = 25^\circ\text{C}$ 以上で使用する場合は、1°Cにつき5mWを減じる

● 推荐動作条件 / Recommended Operating Conditions (Ta=25°C)

Parameter	Symbol	Min.	Typ.	Max.	Unit
電源電圧	V_{CC}	1	1.25	2	V

● 電氣的特性 / Electrical Characteristics ($T_a=25^\circ\text{C}$, $V_{CC}=1.25\text{V}$)

Parameter	Symbol	Min.	Typ.	Max.	Unit	Conditions
粗信号時電流	I _Q	0.5	3	8	mA	—
入力インピーダンス	Z _{IN}	380	540	720	Ω	f _{IN} = 1kHz
入力利得	G _V	30	37	—	dB	V _{IN} = 0.5mV
チャンネルバランス	CB	—	—	2	dB	V _{IN} = 0.5mV
MPX最大出力電圧	V _{OM}	200	—	—	mV _{p-p}	THD ≤ 3%
MPX 38kHzもれ	V _{OO}	—	1	—	mV	粗信号時
バイロット出力電圧	V _{OP}	460	580	—	mV _{p-p}	粗信号時
チャンネルセパレーション	Sep	25	45	—	dB	基準復調器にて
入力換算雑音電圧	V _{NIIN}	—	1	—	μV _{rms}	38kHz停止時 IHF-A
RF部最大出力電圧	V _{O30}	350	600	—	mV _{rms}	—

オーディオ由

1

高麗文書

ROHM

1148

US Patent Number:5,771,441 Issued to Altstatt

Item Number 2: Receiver, TA 7766AF

TOSHIBA

TA7766AF

ELECTRICAL CHARACTERISTICS (Unless otherwise specified, $T_0 = 25^\circ\text{C}$, $V_{CC} = 1.5\text{V}$, $f_m = 1\text{kHz}$)

CHARACTERISTIC	SYMBOL	TEST CIRCUIT	TEST CONDITION	MIN.	typ.	MAX.	UNIT
Supply Current	I_{CC}	—	AC lamp off	—	0.6	1.6	mA
Input Resistance	R_{IN}	—	—	—	36	—	k Ω
Output Resistance	R_{OUE}	—	—	—	19	—	k Ω
Max. Composite Signal Input Voltage	$V_{IN(MAX)}$ (STEREO)	—	$L: R = 50\text{k}\Omega$, $P = 10\text{mW}$, $TED = 5\text{k}\Omega$ $SW_1 = \text{RLED} = 50\text{k}\Omega$ $SW_2 = \text{LPP ON}$	—	250	—	mV _{rms}
Separation	Sep	—	$L: R = 90\text{mVrms}$, $f_m = 10\text{kHz}$ $P = 10\text{mVrms}$, $f_m = 1\text{kHz}$ $SW_1 = \text{RLED} = 50\text{k}\Omega$ $SW_2 = \text{LPP ON}$	—	30	—	dB
Total Harmonic Distortion	Monaural	THD (MONAURAL)	$V_{IN} = 100\text{mVrms}$ $SW_1 = \text{RLED} = 500\Omega$	—	0.2	1.5	—
	Stereo	THD (STEREO)	— $L: R = 90\text{mVrms}$, $P = 10\text{mVrms}$ $SW_1 = \text{RLED} = 50\text{k}\Omega$ $SW_2 = \text{LPP ON}$	—	0.4	—	—
Voltage Gain	G_V	—	$V_{IN} = 100\text{mVrms}$ $SW_1 = \text{RLED} = 500\Omega$	-4	-2	1	dB
Channel Balance	CB	—	$V_{IN} = 100\text{mVrms}$ $SW_1 = \text{RLED} = 500\Omega$	—	0	2.0	dB
Luma ON Sensitivity	$V_L(\text{ON})$	—	Pin ₁ $SW_1 = \text{RLED} = 50\text{k}\Omega$	—	—	5	—
Luma OFF Sensitivity	$V_L(\text{OFF})$	—	Input $SW_1 = \text{RLED} = 500\Omega$	7	—	—	mV _{rms}
Stereo Lamp Hysteresis	V_H	—	To turn-off from turn-on	—	3	—	mV _{rms}
Capture Range	CR	—	$P = 10\text{mVrms}$	—	±3	—	—
Carrier leak (Note)	19kHz	CL	$L: R = 90\text{mVrms}$ $P = 10\text{mVrms}$ $SW_1 = \text{RLED} = 50\text{k}\Omega$	—	30	—	—
	38kHz	CL	— $P = 10\text{mVrms}$, $L: R = 90\text{mVrms}$ $SCA = 10\text{mVrms}$, $f_{SCA} = 67\text{kHz}$ $SW_1 = \text{RLED} = 50\text{k}\Omega$	—	50	—	dB
SCA Rejection Ratio	SCA Rej	—	$P = 10\text{mVrms}$, $L: R = 90\text{mVrms}$ $SCA = 10\text{mVrms}$, $f_{SCA} = 67\text{kHz}$ $SW_1 = \text{RLED} = 50\text{k}\Omega$	—	70	—	dB
Signal to Noise Ratio	S/N	—	$V_{IN} = 100\text{mVrms}$, $R_p = 520\Omega$ $SW_1 = \text{RLED} = 500\Omega$	—	65	—	dB

(Note) Carrier leak of 38kHz is only carrier.

US Patent Number:5,771,441 Issued to Altstatt

Item Number 3: Receiver, TA 7792F

TOSHIBA

TA7792P/F

MAXIMUM RATINGS ($T_a = 25^\circ\text{C}$)

CHARACTERISTIC	SYMBOL	RATING	UNIT
Supply Voltage	V_{cc}	5	V
Power Dissipation	P_D (Note)	750	mW
	TA7792P	350	
Operating Temperature	T_{op}	-25~75	°C
Storage Temperature	T_{sg}	+55~150	°C

(Note) Operated above $T_a = 25^\circ\text{C}$ in the proportion of $6\text{mW}/^\circ\text{C}$ for TA7792P and of $2.8\text{mW}/^\circ\text{C}$ for TA7792F.

ELECTRICAL CHARACTERISTICS

Unless otherwise specified, $T_a = 25^\circ\text{C}$, $V_{cc} = 1.5\text{V}$

FM : $V_{in} = 600\mu\text{V}$ (EMF), $f = 83\text{MHz}$, $f_m = 1\text{kHz}$, $df = \pm 22.5\text{kHz}$
 AM : $V_{in} = 600\mu\text{V}$ (EMF), $f = 1\text{MHz}$, $f_m = 1\text{kHz}$, MOD = 30%

CHARACTERISTIC	SYMBOL	TEST CIRCUIT	TEST CONDITION	MIN.	TYP.	MAX.	UNIT
Supply Current	I_{cc} (FM)	1	$V_{in} = 0$	—	4.0	5.2	mA
	I_{cc} (AM)	1	$V_{in} = 0$	—	1.2	1.8	
Input Limiting Voltage	V_{in} (FM)	1	— 1dB limiting	—	10	16	$\text{dB}/\mu\text{V}$ (EMF)
Total Harmonic Distortion	THD (FM)	1	—	—	0.25	—	%
Signal To Noise Ratio	S/N (FM)	1	—	—	62	—	dB
Quiescent Sensitivity	O_s	1	$S/N = 30\text{dB}$	—	12	—	$\text{dB}/\mu\text{V}$ (EMF)
AM Rejection Ratio	AMR	1	MOD = 30%	—	30	—	dB
Oscillator Voltage	V_{osc}	2	$f = 10\text{MHz}$	5.1	90.1	135	mV/cfs
Oscillator Step Supply Voltage	V_{osc} (FM)	1	$V_{in} < -20\text{dB}/\mu\text{V}$ EMF	—	0.65	0.95	V
Recovered Output Voltage	V_{op} (FM)	1	—	28	49	68	mV/cfs
Voltage Gain	G_V	1	$V_{in} = 30\text{dB}/\mu\text{V}$ EMF	14	26	50	mV/mV
Recovered Output Voltage	V_{op} (AM)	1	—	25	49	69	mV/cfs
Total Harmonic Distortion	THD (AM)	1	—	—	1.5	—	%
Signal To Noise Ratio	S/N (AM)	1	—	—	40	—	dB
Oscillator Step Supply Voltage	V_{osc} (AM)	1	$V_{in} < -20\text{dB}/\mu\text{V}$ EMF	—	0.85	0.95	V
Output Resistance Pin B	R_o (FM)	1	$f = 1\text{kHz}$	—	1.4	—	$\text{k}\Omega$
	R_o (AM)	1	$f = 1\text{kHz}$	—	8	—	

② V_{in} : Open Display

US Patent Number:5,946,343 Issued to Schotz

Item Number 1: Digital Signal Processor, DSP56002

Specifications

DC Electrical Characteristics

DC ELECTRICAL CHARACTERISTICS

Table 2-3 DC Electrical Characteristics

Characteristics	Symbol	Min	Typ	Max	Units
Supply Voltage	V_{CC}	4.5	5.0	5.5	V
Input High Voltage					
• EXTRAL	V_{IH}	4.0	—	V_{CC}	V
• RESET	V_{IHE}	2.5	—	V_{CC}	V
• NODA, MDDA, MDDC	V_{IHM}	2.5	—	V_{CC}	V
• All other inputs	V_{IH}	2.0	—	V_{CC}	V
Input Low Voltage					
• EXTRAL	V_{IL}	-0.5	—	0.6	V
• NODA, MDDA, MDDC	V_{ILM}	-0.5	—	2.0	V
• All other inputs	V_{IL}	-0.5	—	0.8	V
Input Leakage Current	I_{IN}	-1	—	1	μA
External RESET, NODA/IRQA, NODB/IRQB, MDDA/NSA, UP, ER, WT, CKP, PINT, NCZ, NICKLY, MCCLK, D29IN					
Standby (Off-state) Input Current ($\leq 2.4V/0.4V$)	I_{BS}	-10	—	10	μA
Output High Voltage ($I_{OH} = 0.1$ mA)	V_{OH}	2.1	—	—	V
Output Low Voltage ($I_{OL} = 2.0$ mA)	V_{OL}	—	—	0.4	V
IREQ I_{OL} = 0.7 mA, TXD I_{OL} = 0.7 mA					
Internal Supply Current at 40 MHz ¹	I_{CC1} I_{CCW} I_{CCS}	— — —	30 12 2	105	mA
• In Wait mode ²					mA
• In Stop mode ³					mA
Internal Supply Current at 66 MHz ¹	I_{CC1} I_{CCW} I_{CCS}	— — —	95 15 2	130	mA
• In Wait mode ²					mA
• In Stop mode ³					mA
Internal Supply Current at 80 MHz ¹	I_{CC1} I_{CCW} I_{CCS}	— — —	115 18 2	160	mA
• In Wait mode ²					mA
• In Stop mode ³					mA
FLL Supply Current ⁴					
• 40 MHz		—	1.0	1.5	mA
• 66 MHz		—	1.1	1.5	mA
• 80 MHz		—	1.2	1.8	mA
CKOUT Supply Current ⁵					
• 40 MHz		—	14	20	mA
• 66 MHz		—	28	35	mA
• 80 MHz		—	34	42	mA
Input Capacitance ⁶	C_{IN}	—	10	—	μF

Notes: 1. Section 4 Design Considerations describes how to calculate the external supply current.
 2. In order to obtain these results all inputs must be terminated (i.e. not allowed to float).
 3. Values are given for FLL enabled.
 4. Values are given for CKOUT enabled.
 5. Periodically sampled and not 100% tested.

US Patent Number: 5,946,343 Issued to Schotz

Item Number 2: A/D Converter, SAA7360

Philips Semiconductors	Product specification
Bitstream conversion ADC for digital audio systems	SAA7360

Table 1 Output data formats

ODF2	ODF1	MODE
0	0	test
0	1	format 1
1	0	format 2
1	1	PS

Reset:

When the RESET pin is held LOW the data outputs are set to zero. The RESET pin operates as a Schmitt trigger, enabling a power-on reset function by using an external RC circuit.

LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V_{DD}	analog supply voltage	note 1	-0.5	+5.5	V
V_{I}	DC input voltage		-0.5	+5.5	V
I_{SC}	DC input diode current		-	± 10	mA
V_{O}	DC output voltage		-0.5	$V_{DD} + 0.5$	V
I_{O}	DC output source or sink current		-	± 10	mA
$I_{ON,125}$	Total DC V_{DD} or V_{DD} current		-	± 0.5	A
T_{J}	operating ambient temperature		-40	+85	°C
T_{ZS}	storage temperature		-55	-150	°C
V_{SS}	electrostatic discharging	note 2	-2000	-2000	V
		note 3	-200	-200	V

Notes:

1. All V_{DD} and V_{SS} pins must be externally connected to the same power supply.
2. Equivalent to discharging a 100 pF capacitor via a 1.5 kΩ series resistor with a rise time of 15 ns.
3. Equivalent to discharging a 220 pF capacitor via a 1.6 μ H series inductor.

CHARACTERISTICS

$V_{DD} = 5$ V, $T_{J} = 25$ °C, $t_{PD} = 250$ ns, $t_{R} = 44.4$ ns; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supplies						
V_{DD}	analog supply voltage		4.5	5.0	5.5	V
I_{DD}	analog supply current		-	45	-	mA
V_{DD}	digital supply voltage		4.5	5.0	5.5	V
I_{DD}	digital supply current		-	50	-	mA
P_{DD}	total power consumption		-	465	-	mA

US Patent Number:5,946,343 Issued to Schotz

Item Number 3: Stereo Filter MPEG, SAA2520

Philips Semiconductor		Preliminary specification					
Stereo filter and codec for MPEG layer 1 audio applications		SAA2520					
LIMITING VALUES In accordance with the Absolute Maximum System (IEC 134).							
SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.			
V_{DD}	Supply voltage		-0.5	5.5			
V_I	Input voltage	note 6	-0.5	$V_{DD} - 0.5$			
I_{SS}	Supply current from V_{DD}		-	160			
I_{SD}	Supply current to V_{DD}		-	160			
I_S	Input current		-10	10			
I_O	Output current		-20	20			
P_{DS}	Total power dissipation		-	850			
T_{JG}	Storage temperature range		-65	160			
T_{OP}	Operating ambient temperature range		-40	85			
V_{BR}	Electrostatics handling	note 2	-1800	1800			
V_{ES}	Electrostatics handling	note 3	-70	70			

Notes:

1. Input voltage should not exceed 5.5 V unless otherwise specified.
2. Equivalent to discharging a 100 pF capacitor through a 1.8 kΩ series resistor.
3. Equivalent to discharging a 100 pF capacitor through a 0.0 series resistor.

DC CHARACTERISTICS
 $T_{OP} = -40 \text{ to } 85^\circ\text{C}$; $V_{DD} = 3.6 \text{ to } 5.5 \text{ V}$ unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply						
V_{DD}	Supply voltage range		2.0	5.0	5.5	V
I_{CO}	Operating current	$V_{DD} = 3 \text{ V}$ (note 1)	-	62	110	mA
I_{SD}	Operating current	$V_{DD} = 3.6 \text{ V}$ (note 1)	-	28	60	mA
Inputs URD4, SBD4, SBF4, LTC4L, LTC4TO, LTC4T4, X22IN, X24IN						
V_{IL}	HIGH level input voltage		0.7 V_{DD}	-	-	V
V_{IL}	LOW level input voltage		-	-	0.3 V_{DD}	V
\bar{I}_1	Input current	$V_I = 0 \text{ V}$ $T_{OP} = 25^\circ\text{C}$	-	-	10	mA
\bar{I}_1	Input current	$V_I = 5.5 \text{ V}$ $T_{OP} = 25^\circ\text{C}$	-	-	10	mA
Outputs PWDRDN, LTEN4						
V_{IL}	HIGH level output voltage		0.7 V_{DD}	-	-	V
V_{IL}	LOW level output voltage		-	-	0.3 V_{DD}	V
\bar{I}_1	Output current	$V_I = V_{DD}$ $T_{OP} = 25^\circ\text{C}$	40	-	250	mA

US Patent Number:5,946,343 Issued to Schotz

Item Number 4: MPEG, SAA2521

Philips Semiconductors		Preliminary specification				
Masking threshold processor for MPEG layer 1 audio compression applications		SAA2521				
DC CHARACTERISTICS $V_{DD} = 2.0$ to 5.5 V; $T_{A} = -40$ to 85°C (unless otherwise specified)						
SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply						
V_{DD}	Supply voltage range		3.0	3	3.5	V
I_{DD}	Operating current	$V_{DD} = 3.0$ V	-	10	20	mA
I_{DD}	Operating current	$V_{DD} = 5$ V	-	50	60	mA
I_{DDON}	Start-up current	In power-down mode	-	100	-	mA
Input						
V_{IN}	Low level input voltage		0	-	$0.2V_{DD}$	V
V_{IN}	High level input voltage		$0.7V_{DD}$	-	V_{DD}	V
I	Input current		-	-	10	mA
Output						
V_{OUT}	Low level output voltage	Node 1	-	-	0.4	V
V_{OUT}	High level output voltage	Node 1	$V_{DD} - 0.5$	-	-	V
2-state outputs						
I_{OFF}	OFF state current	$V_{I} = 0$ to 6.5 V	-	-	10	mA

Note

1. Maximum load current for LTCDATA, LTCNTIC, LTCNTOC, LTCNG, LTCOLIC, TGT1, TGT2, FDAC, FCAP = 2 mA;
for LTCDATA = 3 mA.

US Patent Number:5,946,343 Issued to Schotz

Item Number 5: Modulator, RF2422

RF2422

Absolute Maximum Ratings

Parameter	Rating	Unit
Supply Voltage	-0.5 to +7.5	V _{CC}
Input DC and RF Levels	+10	dBm
Operating Ambient Temperature	-40 to +85	°C
Storage Temperature	-40 to +150	°C

Caution: ESD sensitive device.

DO NOT EXPOSE THIS DEVICE TO TEMPERATURES OUTSIDE THE OPERATING RANGE FOR LONG PERIODS OF TIME. THIS CAN DAMAGE THE DEVICE. DO NOT EXPOSE THIS DEVICE TO TEMPERATURES OUTSIDE THE OPERATING RANGE FOR LONG PERIODS OF TIME. THIS CAN DAMAGE THE DEVICE.

Parameter	Specification			Unit	Condition
	Min.	Typ.	Max.		
Carrier Input Frequency Range	600		2500	MHz	T=25°C, V _{CC} =5V
Power Level Input VSWR	-4	5.1	-6	dBm	At 100MHz At 1500MHz At 2500MHz
Modulation Input Frequency Range	20	30	50	MHz	
Reference Voltage (Vref)	2.0		Vref±1.0	V	
Maximum Modulation (MSK)				V	
Carrier Attenuation	0.2		0.3	dB	
Quadrature Phase Error	3		4	°	
Input Bias Current	20		40	mA	
RF Output	-3		-3	dBm	LO=2GHz and -5dBm, IQD=2.0Vpp, SSB
Output Impedance	50			Ω	
Output VSWR	3.0:1		1.5:1		At 100MHz At 1500MHz At 2500MHz
Reference Output	40	-25		dBm	
Carrier Suppression	25	45		dB	
IF Suppression	20	45		dB	
Ground Noise Floor	25	120	-145	dBmHz	Termination of the carrier and the carrier RF signal Termination of local oscillator signals At 20MHz offset, V _{CC} =5V Tied to Vcc (10, 20, 30, 40, and 50MHz) At 250MHz At 1500MHz
Power Down			100	dB	
Turn On/Off Time	50		2.5	μs	
PD Input Resistance			10	V	
Power Down (PD)	1.0	1.2		V	
Power Down (GFP)				V	
Power Supply Voltage	4.5	9	6.0	V	Specification
Current		45	50	mA	Operating time Operating current Operating power
			55	64	Power down

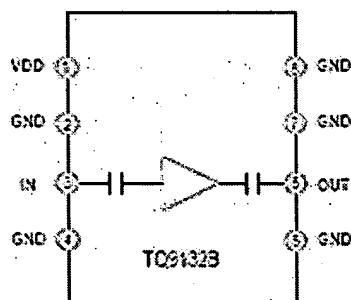
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5-10

Rev A8-010817

US Patent Number:5,946,343 Issued to Schotz

Item Number 6: Power Amplifier, TQ9132



Product Description

The TQ9132B amplifier is an 800-2500 MHz amplifier capable of providing moderate output power (50 mW) for a wide variety of transmit and receive applications. The amplifier's input and output are matched to 50 Ω with internal circuitry, simplifying interfaces to 50 Ω systems. In addition, DC blocking capacitors are included on chip, permitting direct connections to the input and output. Its 8-pin surface-mount package and low cost are well suited to many wireless communications applications.

Electrical Specifications¹

Parameter	Min	Typ	Max	Units
Gain	13.5	15	16	dB
Output 5 dB Gain Compression	15.5	17	—	dBm
Input Return Loss	—	12	—	dB
Output Return Loss	—	12	—	dB
DC Supply Current	65	120	140	mA

Note 1: Test Condition: V_{DD} = 5.0 V, Freq. = 1500 MHz, T = 27°C.
Note 2: MSOP package 500 mils thick.

TQ9132B

DATA SHEET

3V Cellular TDMA/AMPS Power Amplifier IC

Features

- Single 3V-5V supply
- Wide frequency range
- +15 dBm output power
- Input and output matched to 50 Ω
- 808 surface-mount plastic package

Applications

- Power Amplifier drivers
- PON/Medium-power amplifiers
- Medium-power WLANs
- CDMA Modems
- Base Station receivers

For additional information and direct quotations, see our website: www.triquint.com.

US Patent Number:5,946,343 Issued to Schotz

Item Number 7: Phase Locked Loop, MC12210

MC12210

ELECTRICAL CHARACTERISTICS ($V_{DD} = 0.3 \text{ to } 6.5 \text{ V}$; $T_A = -40 \text{ to } +85^\circ\text{C}$ unless otherwise noted)

Parameter	Symbol	Min	Typ	Max	Unit	Condition
Supply Current for V_{DD}	I_{DD}	—	2.5	130	mA	Note 1
		—	10.2	180	mA	Note 2
Supply Current for V_P	I_P	—	0.7	1.1	mA	Note 3
		—	0.6	1.3	mA	Note 4
Operating Frequency (Internal)	f_{PLL}	2500	—	—	MHz	Note 5
Operating Frequency (V_{DDC})	f_{OSC}	—	12	30	MHz	Crystal Mode
		—	—	40	MHz	External Reference Mode
Input Sensitivity	I_{IN}	910	—	1000	dBm	
	$GSIN$	900	—	2000	dBm	
Input HIGH Voltage	V_{IH} (CLK, DATA, LE, FC)	$0.7 V_{DD}$	—	—	V	
Input LOW Voltage	V_{IL} (CLK, DATA, LE, FC)	V_L	—	$0.3 V_{DD}$	V	$V_{DD} = 5.5 \text{ V}$
Input HIGH Current (DATA and CLK)	I_{IH}	—	1.0	2.0	mA	$V_{DD} = 5.5 \text{ V}$
Input LOW Current (DATA and CLK)	I_{IL}	—	—0.5	—	mA	$V_{DD} = 5.5 \text{ V}$
Input Output (OSC)	I_{OSC}	—	150	—	mA	$OSC_{IN} = V_{DD}$ $OSC_{IN} = V_{DD} - 2.2 \text{ V}$
Input HIGH Current (LE and FC)	I_{IH}	—	1.0	2.0	mA	
Input LOW Current (LE and FC)	I_{IL}	—0.5	—0.5	—	mA	
Charge Pump Output Current	I_{QSP} (5V)	—0.8	—2.0	—4.4	mA	$V_{DD} = V_{DD2}, V_P = 2.7 \text{ V}$
Ground SW	I_{GSW} (5V)	—1.4	—2.0	—2.8	mA	$V_{DD} = V_{DD2}, V_P = 2.7 \text{ V}$
	I_{GSW} (2V)	—0.5	—	—1.6	mA	$0.5 \cdot V_{DD} \cdot V_P - 0.5$ $0.5 \cdot V_{DD2} \cdot V_P - 0.5$
Output HIGH Voltage (LD, OR, OR, TOUT)	V_{OH}	4.4	—	—	V	$V_{DD} = 5.5 \text{ V}$
Output LOW Voltage (LD, OR, OR, TOUT)	V_{OL}	—0.4	—	—	V	$V_{DD} = 5.5 \text{ V}$
		—	—	0.4	V	$V_{DD} = 5.5 \text{ V}$
Output HIGH Current (LD, OR, OR, TOUT)	I_{OH}	—1.0	—	—	mA	
Output LOW Current (LD, OR, OR, TOUT)	I_{OL}	1.0	—	—	mA	

1. $V_{DD} = 3.3 \text{ V}$, all outputs open
2. $V_{DD} = 5.5 \text{ V}$, all outputs open
3. $V_{DD} = 3.3 \text{ V}$, all outputs open

4. $V_P = 0.6 \text{ V}$, all outputs open
5. AC coupling, f_{PLL} measured with a 1000 pF capacitor
6. Subsequent numbers and shapes are not to scale, see drawings

Figure 8. Typical External Charge Pump Circuit

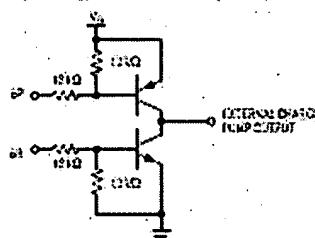
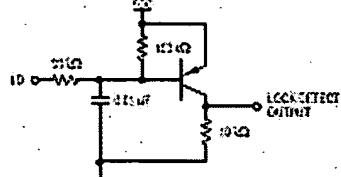


Figure 9. Typical Lock Detect Circuit



US Patent Number:5,946,343 Issued to Schotz

Item Number 8: Voltage Controlled Oscillator, SMV2500

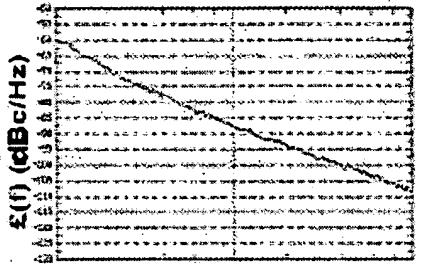


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SMV2500L
VOLTAGE CONTROLLED OSCILLATOR



PHASE NOISE (1 Hz BW, typical)



The graph shows Phase Noise (dBc/Hz) on the y-axis (ranging from -120 to -20) versus Offset (Hz) on the x-axis (logarithmic scale from 10^0 to 10^3). Two curves are plotted: one for 2400 MHz and one for 2454 MHz. Both curves show a minimum phase noise around 100 Hz offset, with the 2454 MHz curve being slightly lower than the 2400 MHz curve at higher offsets.

FEATURES

- Frequency Range: 2400 - 2454 MHz
- Tuning Voltage: 0-3 Vdc
- SUS-5 - Style Package

APPLICATIONS

- Personal Communications Systems
- WLAN
- Portable Radios

OPERATING AND SPECIFICATIONS

	VALUE	UNITS
Oscillation Frequency Range	2400 - 2454	MHz
Phase Noise @ 13 kHz offset (1 Hz BW, typ.)	-37	dBc/Hz
Harmonic Suppression (std. typ.)	-20	dB
Tuning Voltage	0-3	Vdc
Tuning Sensitivity (avg.)	105	MHz/V
Power Output	9.25±2.75	dBm
Load Impedance	50	Ω
Input Capacitance (max.)	50	pF
Pushing	-30	MHz
Pulling (14 dB Return Loss, Any Phase)	-43	MHz
Operating Temperature Range	-40 to 85	°C
Package Style	SUS-5	

POWER SUPPLY REQUIREMENTS

	VALUE	UNITS
Supply Voltage (Vcc, nom.)	3	Vdc
Supply Current (Icc, typ.)	19	mA

All specifications are typical unless otherwise noted and subject to change without notice.

APPLICATION NOTES

- AN-1201: Mounting and Grounding of VCOs
- AN-122: Proper Output Loading of VCOs
- AN-107: How to Solder Z-COMM VCOs

NOTES:

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Rev.1

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US Patent Number:5,946,343 Issued to Schotz

Item Number 9: Low Noise Amplifier, MBA86576

Absolute Maximum Ratings

Symbol	Parameter	Units	Absolute Maximum
V_g	Device Voltage, RF output to ground	V	9
V_s	Device Voltage, RF Input to ground	V	-10
P_{in}	CW RF Input Power	dBm	+13
T_{ch}	Channel Temperature	°C	100
T_{stg}	Storage Temperature	°C	-65 to 150

Thermal Resistance¹⁾
 $\theta_{th} = 110\text{ °C/W}$

Notes:
1. Operation of this device above any one of these limits may cause permanent damage.
2. $T_j = 100\text{ °C}$ (T_c is defined to be the temperature at the package pins where contact is made to the die on board).

MGA86576 Electrical Specifications, $T_c = 25\text{ °C}$, $Z_o = 50\text{ Ω}$, $V_s = 5\text{ V}$

Symbol	Parameters and Test Conditions	Units	Min.	Typ.	Max.
G_p	Power Gain ($ S_{21} ^2$)	dB	-50	21.2	
	$f = 1.5\text{ GHz}$			21.5	
	$f = 2.6\text{ GHz}$			22.1	
	$f = 4.0\text{ GHz}$			19.8	
	$f = 6.0\text{ GHz}$			15.1	
NF_{zo}	60Ω Noise Figure	dB		22	
	$f = 1.5\text{ GHz}$			19	
	$f = 2.6\text{ GHz}$			20	
	$f = 4.0\text{ GHz}$			23	
	$f = 6.0\text{ GHz}$			25	
NF_o	Optimum Noise Figure (Input tuned for lowest noise figure)	dB		16	
	$f = 1.5\text{ GHz}$			15	
	$f = 2.6\text{ GHz}$			15	
	$f = 4.0\text{ GHz}$			19	
	$f = 6.0\text{ GHz}$			21	
P_{1dB}	Output Power at 1 dB Gain Compression	dBm		64	
	$f = 1.5\text{ GHz}$			70	
	$f = 2.6\text{ GHz}$			68	
	$f = 4.0\text{ GHz}$			48	
	$f = 6.0\text{ GHz}$			93	
IP_3	Third Order Intercept Point	dBm		16.0	
$VSWR$	Input VSWR			2.1	
	$f = 1.5\text{ GHz}$			2.1	
	$f = 2.6\text{ GHz}$			2.1	
	$f = 4.0\text{ GHz}$			1.81	
	$f = 6.0\text{ GHz}$			1.21	
	Output VSWR			2.1	
	$f = 1.5\text{ GHz}$			2.1	
	$f = 2.6\text{ GHz}$			1.21	
	$f = 4.0\text{ GHz}$			1.61	
	$f = 6.0\text{ GHz}$			1.21	
I_g	Device Current	mA	9	10	12

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Item Number 10: Digital Interface Transmitter, CS8402

CHSSTEL

CS8401A CS8402A

ABSOLUTE MAXIMUM RATINGS (GND = 0V, all voltages with respect to ground.)

Parameter	Symbol	Min	Max	Units
DC Power Supply	V _{D+}		6.0	V
Input Current, Any Pin Except GND	I _{IN}	-	±10	mA
Digital Input Voltage	V _{INP}	-0.3	V _{D+}	V
Ambient Operating Temperature (power applied)	T _A	-45	125	°C
Storage Temperature	T _{STG}	-45	100	°C

Note: 1. Transient currents of up to 100 mA will not cause SCR latchup.

WARNING: Operation at or beyond these limits may result in permanent damage to the device.
Normal operation is not guaranteed at these extremes.

RECOMMENDED OPERATING CONDITIONS (GND = 0V, all voltages with respect to ground)

Parameter	Symbol	Min	Typ	Max	Units
DC Voltage	V _{D+}	4.5	5.0	6.0	V
GND Current	I _{GND}	-	1.0	5	mA
Ambient Operating Temperature: CS8401A/CP or 'CS Note 3	T _A	0	25	70	°C
CS8401A/P or 'IS		-40	45	85	°C
Power Consumption	P _D	-	7.5	25	mw

Note: 2. Drivers open (unbiased); the majority of power is used by the load connected to the drivers.
3. The 'CP' and 'IS' parts are specified to operate over 0 to 70 °C but are tested at 25 °C only.
The 'P' and 'I' parts are tested over the full -40 to 85 °C temperature range.

DIGITAL CHARACTERISTICS

(TA = 25 °C for suffixes 'CP' & 'CS', TA = -40 to 85 °C for 'IP' & 'IS'; V_{D+} = 5V ± 10%)

Parameter	Symbol	Min	Typ	Max	Units
High-Level Input Voltage	V _{IH}	2.0	V _{D+} -0.3	V	
Low-Level Input Voltage	V _{IL}	-0.3	-0.6	V	
High-Level Output Voltage (I _O = 200mA)	V _{OH}	V _{D+} -0.3		V	
Low-Level Output Voltage (I _O = 3.0mA)	V _{OL}	-	0.4	V	
Input Leakage Current	I _{IN}	-	1.0	10	mA
Master Clock Frequency: CS8401A Note 4	MCK	-	22	MHz	
CS8402A Note 4		-	7.5	MHz	
Master Clock Duty Cycle: CS8401A	-	40	60	%	

Note: 4. MCK for the CS8401 must be 128, 192, 256, or 384x the input word rate (based on M2 and M3 to control register 2). MCK for the CS8402A must be 128x the input word rate, except in Transparent mode where MCK is 256x the input word rate.

Specifications are subject to change without notice.

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Item Number 11: Digital to Analog Converter, TDA1305T

Philips Semiconductors		Preliminary specification					
Stereo 1fs data input up-sampling filter with bitstream continuous dual DAC (SCC-DAC2)				TDA1305T			
CLOCK REFERENCE DATA							
SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.		
V_{DDO}	digital supply voltage	note 1	3.2	5.0	6.5		
V_{A1}	analog supply voltage	note 1	2.4	5.0	6.5		
V_{DDA}	operational amplifier supply voltage	note 1	2.4	5.0	6.5		
I_{DD}	digital supply current	$V_{DDO} = 5\text{ V}$ at code 00000H	-	33	-		
I_{AA}	analog supply current	$V_{DDO} = 5\text{ V}$ at code 00000H	-	5.5	9		
I_{OA}	operating amplifier supply current	$V_{DDO} = 5\text{ V}$ at code 00000H	-	6.5	9		
V_{DOUT}	full-scale output voltage (PMS value)	$V_{DDO} = V_{A1} = V_{DDA} = 5\text{ V}$	1.425	1.5	1.575		
$(THD + N)_{DS}$	total harmonic distortion SUS noise-to-signal ratio	at 0 dB signal level	-	-90	-31		
			-	0.003	0.006		
		at -20 dB signal level	-	-14	-10		
			-	0.65	0.1		
		at -60 dB signal level; A-weighted	-	-43	-		
SN	signal-to-noise ratio at 0dBFS	A-weighted; at code 00000H	100	103	-		
BR _{DS}	input bit rate at data input	$f_s = 48\text{ kHz}$; normal speed	-	-	3.072		
BR _{DS}	input bit rate at data input	$f_s = 48\text{ kHz}$; double speed	-	-	6.144		
f_s	system clock frequency	-	6.4	-	98.432		
TC _{DS}	full scale temperature coefficient at analog outputs (VOL and VOR)	-	-	2100×10^{-6}	-		
T_{A1}	operating ambient temperature	-	-30	-	+85		

Note

1. All V_{DD} and V_{A1} pins must be connected to the same supply.

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Item Number 12: Clock Recovery & Timing, TRU-050

Parameter	Symbol	Min	Max	Unit
Input Voltage Range	U _{IN}	0.003	65.516	Vdc
Output Voltage Range	U _{OUT}	0.003	37.463	Vdc
Normal Output Frequency				
Angle 1	θ ₁	120	135.50	°
Angle 2	θ ₂	0.05	32.363	°
Supply Voltage	U _{sup}	45	55	V
Supply Current (Typ. ±35%)	I _{sup}	75	43	mA
Output Voltage (Typ. ±45%)	U _{OUT}	0	37.463	Vdc
Angle Frequency ¹	θ ₁ freq	7.5	-	°/s
Angle Frequency ²	θ ₂ freq	-	0.5	°/s
Position Angle ¹				
Position (0.0 To 2.51)	θ ₁ pos	0.0	5	°
Position (25.0 To 0.51)	θ ₂ pos	0.5	5	°
Supply Angle ²				
Angle 1	θ ₁ ang	0	60	°
Angle 2	θ ₂ ang	0	35	°
Position Gain	G _{pos}	40	60	°
Position Bias				
Position 1	θ ₁ posb	20	-	°
Position 2	θ ₂ posb	-	0.0	°
Supply Angle Gain (Typ. ±25%)	G _{ang}	50	-	°/s
Supply Angle (Typ. ±10%)	θ ₁ angtyp	50	100	°
Position Frequency ¹				
Angle Position	θ ₁ freq	-	-	°/s
Position Gain	G _{posfreq}	25	-	°/s
Position Bias	θ ₁ posfreq	-	0.5	°
Normal Output Frequency (Typ. ±35%)				
Angle 1	θ ₁ freq	75.001	75.001	°/s (positive)
Angle 2	θ ₂ freq	-75.001	75.001	°/s (positive)
Position Gain (typ)	G _{pos}	40	60	°/s

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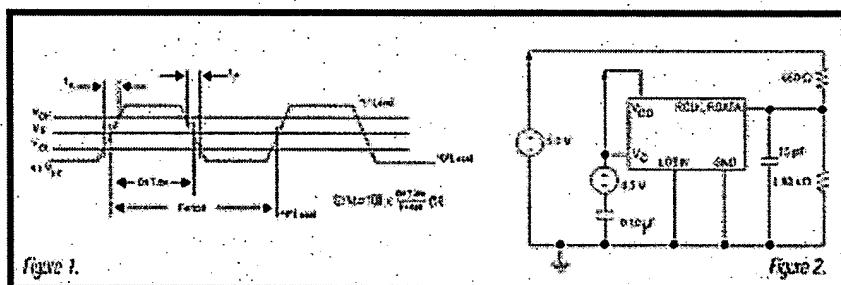


figure 1

Figure 2.

US Patent Number: 5,946,343 Issued to Schotz

Item Number 13: Demodulator, RF2703

RF2703

Absolute Maximum Ratings

Parameter	Rating	Unit
Supply Voltage	-0.5 to 7.0	V _{cc}
Input Level	500	mV _{pp}
Operating Ambient Temperature	-40 to +85	°C
Storage Temperature	-40 to +150	°C



Caution ESD sensitive device.

DO NOT Ground before the following test has been completed in accordance with the test of the patent. (However, RF2703 Demodulator may be safely discharged to ground after a short time. (See the General section for more information on the use of the circuit and products.)

Parameter	Specification			Unit	Condition
	Min.	Typ.	Max.		
Overall IF Frequency Range		0.1 to 250		MHz	T=25°C, V _{cc} =3.0V, IF=100MHz, LO=200MHz, F _{IF} =500Hz. For IF frequencies below ~2.5MHz, the LO should be a square wave. IF frequencies lower than 100Hz are obtained if the LO is a square wave and sufficiently large DC blocking capacitors are used.
Baseband Frequency Range	DC to 50			MHz	
Input Impedance	(200 ± 1%)			Ω	Each input, single-ended
LO					
Frequency					Twice the IF frequency. For IF frequencies below ~2.5MHz, the LO should be a square wave. IF frequencies lower than 100Hz are obtained if the LO is a square wave and sufficiently large DC blocking capacitors are used.
Level	0.05 to 1			V _{pp}	
Input Impedance	(500 ± 1%)			Ω	
Demodulator Configuration					V _{IF} =2.5mV _{pp} , LO=230mV _{pp} , Z _{IF} =10kΩ
Output Impedance					Each output, low and GND
Maximum Output Voltage Gain	50 (± 10%)	1.4	20	V _{pp}	Saturated
Noise Figure	22.5	26	29.1	dB	V _{cc} =3.0V
Input Third Order Intercept Point (I _{IP3})	-12	-24	-35	dBm	V _{cc} =5.0V
	-11	-23	-34	dBm	Single Sideband, IF Input of device respectively matched
	-10	-22	-33	dBm	Single Sideband, GND short resistor at IF Input
	-9	-21	-32	dBm	V _{cc} =3.0V, IF Input of device respectively matched
	-8	-20	-31	dBm	V _{cc} =3.0V, GND short resistor at IF Input
	-7	-19	-30	dBm	V _{cc} =5.0V, IF Input of device respectively matched
	-6	-18	-29	dBm	V _{cc} =5.0V, GND short resistor at IF Input
	-5	-17	-28	dBm	V _{cc} =5.0V, IF Input of device respectively matched, Z _{IF} =50Ω
LO Amplitude Balance, Quadrature Phase Error	0.1	0.5	dB		V _{cc} =3.0V, I _{out} and Q _{out} to GND
DC Output	±0.0	±0.0	mA		V _{cc} =3.0V, I _{out} and Q _{out} to GND
DC Offset	±0.0	±2.4	2.5	V	I _{out} to Q _{out}
	±10	±50	μV		

7

QUADRATURE
DEMODULATORS

US Patent Number:5,946,343 Issued to Schotz

Item Number 13: Demodulator, RF2703 continued

RF2703

Modulator Configuration					$I_{F1}=24\text{mA}_{\text{DC}}$, $I_{L1}=200\text{mA}_{\text{DC}}$, $Z_{\text{load}}=125\text{Ω}$ Saturated Single Sideband, 1dB Gain Compression, Single Sideband
Maximum Output	200				
Input Voltage	90				
Voltage Gain	6				
dB Amplitude Distortion	0.1				
Quadrature Phase Error	±1				
Carrier Suppression	25				
Sideband Suppression	90				Unadjusted. Carrier Suppression may be optimized further by adjusting the DC offset level between the A and B inputs.
Power Supply					Operating limits $V_{cc}=3.0\text{V}$ $V_{cc}=5.0\text{V}$
Voltage	2.7 to 6				
Current	8	10	12		mA

7

QUADRATURE
DEMODULATORS

US Patent Number:5,946,343 Issued to Schotz

Item Number 14: Microprocessor, PIC16C55

PIC16C5X

12.1 DC Characteristics: PIC16C54/55/56/57-RC, XT, 10, HS, LP (Commercial)

PIC16C44/56/58/57-RC, XT, 10, HS, LP (Commercial)			Standard Operating Conditions (unless otherwise specified) Operating Temperature: 0°C to 70°C (for commercial)				
Param. No.	Symbol	Characteristics/Device	Min	Typ	Max	Units	Conditions
CCC1	V _{DD}	Supply Voltage PIC16C5X-RC PIC16C5X-XT PIC16C5X-10 PIC16C5X-HS PIC16C5X-LP	3.0	—	3.3	V	
CCC2	V _{DD}	RAM Data Retention Voltage ⁽¹⁾	—	4.5 ⁽²⁾	—	V	Device in SLEEP mode
CCC3	V _{DDR}	V _{DD} Start Voltage to ensure Power-on Reset	—	V _{DD}	—	V	See Section 6.1 for details on Power-on Reset
CCC4	V _{DD}	V _{DD} Rise Rate to ensure Power-on Reset	6.0E ⁽³⁾	—	—	1/Vms	See Section 6.1 for details on Power-on Reset
CCC5	I _{DD}	Supply Current ⁽⁴⁾ PIC16C5X-RC ⁽⁵⁾ PIC16C5X-XT PIC16C5X-10 PIC16C5X-HS PIC16C5X-LP	—	3.6 3.4 4.6 4.6 6.0 15	3.8 3.5 10 10 20 22	mA	Freq = 4 MHz, V _{DD} = 5.5V Freq = 4 MHz, V _{DD} = 6.0V Freq = 10 MHz, V _{DD} = 5.5V Freq = 10 MHz, V _{DD} = 6.0V Freq = 20 MHz, V _{DD} = 5.5V Freq = 20 MHz, V _{DD} = 6.0V V _{DD} = 1.0V, V _{DDT} disabled V _{DD} = 1.0V, V _{DDT} enabled
CCC6	I _{DD}	Power-down Current ⁽⁶⁾	—	4.0 4.6	12 9	mA	V _{DD} = 1.0V, V _{DDT} enabled V _{DD} = 1.0V, V _{DDT} disabled

⁽¹⁾ These parameters are characterized but not tested.

⁽²⁾ Certain "Typ" values are based on characterization results at 25°C. This data is for design guidance only and is not tested.

⁽³⁾ Note: t_{ris} is the time to which V_{DD} can be lowered in SLEEP mode without losing RAM data.

⁽⁴⁾ The supply current is mainly a function of the operating voltage and frequency. Other factors such as bus loading, oscillator type, bus rate, internal code execution pattern and temperature also have an impact on the current consumption.

⁽⁵⁾ a) The test conditions for all I_{DD} measurements (active/operation mode and OSC1 = external square wave, turn off R_{XT} at 800 ns, instruction called to V_{DD}, T_{CKH} = V_{DD}, V_{DDT} = V_{DD}, V_{DDT} enabled/disabled as specified).

⁽⁶⁾ b) For standby current measurements, the conditions are the same, except that the device is in SLEEP mode. The power-down current in SLEEP mode does not depend on the oscillator type.

⁽⁷⁾ Does not include current through R_{XT}. The current through the register can be estimated by the formula: I_{DD} = V_{DD}I_{DDR}/(5A) with R_{XT} in 1Ω.

US Patent Number:5,946,343 Issued to Schotz

Item Number 15: DSSS Transmitter, CYLINK SSTX

NO DATASHEET

US Patent Number:5,946,343 Issued to Schotz

Item Number 16: DSSS Receiver, CYLINK Part# SPECTRE

NO DATASHEET

US Patent Number:5,946,343 Issued to Schotz

Item Number 17: Mixer, IAM81008

NO DATASHEET

US Patent Number:5,946,343 Issued to Schotz

Item Number 18: Channel Encoder/Decoder, SRT241203

NO DATASHEET

US Patent Number:5,946,343 Issued to Schotz

Item Number 19: Interleaver/De-interleaver, SRT-24INT

NO DATASHEET

US Patent Number:5,946,343 Issued to Schotz

Item Number 20: Optical Digital Receiver, HK-3131-01

NO DATASHEET

US Patent Number:5,946,343 Issued to Schotz

Item Number 21: Optical Digital Transmitter, HK-3131-03

NO DATASHEET

US Patent Number:5,946,343 Issued to Schotz

Item Number 22: Voltage Controlled Oscillator, M2 D300

NO DATASHEET

EXHIBIT C

NOTE : A=Altstatt S=Schotz FHSS=Frequency Hopping Spread Spectrum w=with Tx=transmitter

System	Part	Supply Current (in mA)	Size (in inches)	Playtime	Note
					Altstatt's Tx
A(Tx)	BA1404	3	18-pin 0.44 x 0.30		FM Stereo Transmitter
				16+ hours	Tx continuous operation time
S(Tx w SS)	DSP56002	90	144-pin 0.78 x 0.78		Schotz FHSS Tx
	>PLL	1	N/A		PLL located inside DSP56002
	>ckout	14	N/A		ckout located inside DSP56002
	SAA7360		44-pin 0.50 x 0.50		A/D converter
	>analog	43			function of the A/D converter
	>digital	50			function of the A/D converter
	SAA2520	82	44-pin 0.55 x 0.55		Stereo Filter MPEG
	SAA2521	25	44-pin 0.55 x 0.55		MPEG
	RF2422	45	16-pin 0.39 x 0.24		Modulator
	TQ9132	85	8-pin 0.19 x 0.23		Power Amp
	MC12210	10.2	16-pin 0.39 x 0.24		PLL
	SMV2500	19	12-pin 0.28 x 0.28		VCO
	HK-3131-01	no data	no data		Optical Digital Rcvr (*)
	M2 D300	no data	no data		VCO (*)
	SRT241203	no data	no data		FEC (*)
	SRT-24INT	no data	no data		Interleaver (*)
				0.1 hours or 6+ minutes	
A(Tx) equation in hours:					
$\{(60 \times 50 \text{mA-min}) / [(60 \text{ minutes/hour} \times 24 \text{ hour/day})(3 \text{mA})]\} \times (24 \text{hour/day}) = 16.6 \text{ hours}$					
S(Tx w SS) equation in hours:					
$\{(60 \times 50 \text{mA-min}) / [(60 \text{ min./hr} \times 24 \text{ hr/day})(90+1+14+43+50+82+25+45+85+10.2+19 \text{mA})]\} \times (24 \text{hr/day}) = 6.4 \text{ min}$					
where min = minutes and hr = hours					
(*) = Unable to locate datasheet for integrated chip (IC) referenced by Schotz					

NOTE : A=Altstatt S=Schotz FHSS=Frequency Hopping Spread Spectrum w=with Rx=Receiver

System	Part	Supply Current (in mA)	Size (in inches)	Playtime	Note
					Altstatt's Rx
A(Rx)	TA7792	4	16-pin 0.77 x 0.30		AM/FM Tuner System
	TA7766A	0.8	18-pin 0.44 x 0.30		FM PLL
				10+ hours	Rx continuous operation time
S(Rx w SS)	DSP56002	90	144-pin 0.78 x 0.78		Schotz FHSS Rx
	>PLL	1	N/A		PLL located inside DSP56002
	>ckout	14	N/A		ckout located inside DSP56002
	MGA86576	16	4-pin 0.20 x 0.07		LNA
	HK-3131-03	no data	no data		Optical Digital Tx (*)
	CS8402	1.5	28-pin 1.20 x 0.20		Digital Interface Tx
	SAA2520	82	44-pin 0.55 x 0.55		Stereo Filter MPEG
	TDA1305T	42	28-pin 0.70 x 0.40		DAC
	TRU-050	63	16-pin 0.80 x 0.30		Clock Recovery and Timing
	RF2703	10	14-pin 0.34 x 0.24		Demodulator
	MC12210	10.2	16-pin 0.39 x 0.24		PLL
	SMV2500	19	12-pin 0.28 x 0.28		VCO
	SRT241203	no data	no data		FEC (*)
	SRT-24INT	no data	no data		De-interleaver (*)
	IAM81008	no data	no data		Mixer (*)
				0.14 hours or 8+ minutes	
A(Rx) equation in hours:					
{(60x50mA-minutes)/[(60 minutes/hour x 24 hour/day)(4.8mA)]} x (24hour/day)					
S(Rx w SS) equation in hours:					
{(60x50mA-minutes)/[(60 minutes/hour x 24 hour/day)(sum of IC currents in mA)]} x (24hour/day)					
(*) = Unable to locate datasheet for integrated chip (IC) referenced by Schotz					

NOTE : A=Altstatt S=Schotz DSSS=Direct Sequence Spread Spectrum w=with Tx=transmitter

System	Part	Supply Current (in mA)	Size (in inches)	Playtime	Note
					Altstatt's Tx
A(Tx)	BA1404	3	18-pin 0.44 x 0.30		FM Stereo Transmitter
				16+ hours	Tx continuous operation time
S(Tx w SS)	DSP56002	90	144-pin 0.78 x 0.78		Schotz DSSS Tx
	>PLL	1	N/A		PLL located inside DSP56002
	>ckout	14	N/A		ckout located inside DSP56002
	PIC16C55	1.8	28-pin 1.5 x 0.50		Microprocessor
	SAA7360		44-pin 0.50 x 0.50		A/D converter
	>analog	43			function of the A/D converter
	>digital	50			function of the A/D converter
	RF2422	45	16-pin 0.39 x 0.24		Modulator
	MC12210	10.2	16-pin 0.39 x 0.24		PLL
	SMV2500	19	12-pin 0.28 x 0.28		VCO
	CYLINK SSTS	no data	no data		DSSS Transmitter (*)
	HK-3131-01	no data	no data		Optical Digital Rcvr (*)
	M2 D300	no data	no data		VCO (*)
				0.18 hours or 11 minutes	
A(Tx) equation in hours:					
$\{(60 \times 50 \text{mA-minutes}) / [(60 \text{ minutes/hour} \times 24 \text{ hour/day}) (3 \text{mA})]\} \times (24 \text{hour/day})$					
S(Tx w SS) equation in hours:					
$\{(60 \times 50 \text{mA-minutes}) / [(60 \text{ minutes/hour} \times 24 \text{ hour/day}) (\text{sum of IC currents in mA})]\} \times (24 \text{hour/day})$					
(*) = Unable to locate datasheet for integrated chip (IC) referenced by Schotz					

NOTE : A=Altstatt S=Schotz DSSS=Direct Sequence Spread Spectrum w=with Rx=Receiver

System	Part	Supply Current (in mA)	Size (in inches)	Playtime	Note
					Altstatt's Rx
A(Rx)	TA7792	4	16-pin 0.77 x 0.30		AM/FM Tuner System
	TA7766A	0.8	18-pin 0.44 x 0.30		FM PLL
				10+ hours	Rx continuous operation time

S(Rx w SS)	DSP56002	90	144-pin 0.78 x 0.78		Schotz DSSS Rx
	>PLL	1	N/A		PLL located inside DSP56002
	>ckout	14	N/A		ckout located inside DSP56002
	PIC16C55	1.8	28-pin 1.5 x 0.50		Microprocessor
	CYLINK	no data	no data		DSSS Receiver
	MGA86576	16	4-pin 0.20 x 0.07		LNA
	IAM81008	no data	no data		Mixer (*)
	CS8402	1.5	28-pin 1.20 x 0.20		Digital Interface Tx
	TDA1305T	42	28-pin 0.70 x 0.40		DAC
	MC12210	10.2	16-pin 0.39 x 0.24		PLL
	SMV2500	19	12-pin 0.28 x 0.28		VCO
	HK-3131-03	no data	no data		Optical Digital Tx (*)
				0.25 hours or 15 minutes	

A(Rx) equation in hours:

$$\{(60 \times 50 \text{mA-minutes}) / [(60 \text{ minutes/hour} \times 24 \text{ hour/day})(4.8 \text{mA})]\} \times (24 \text{hour/day})$$

S(Rx w SS) equation in hours:

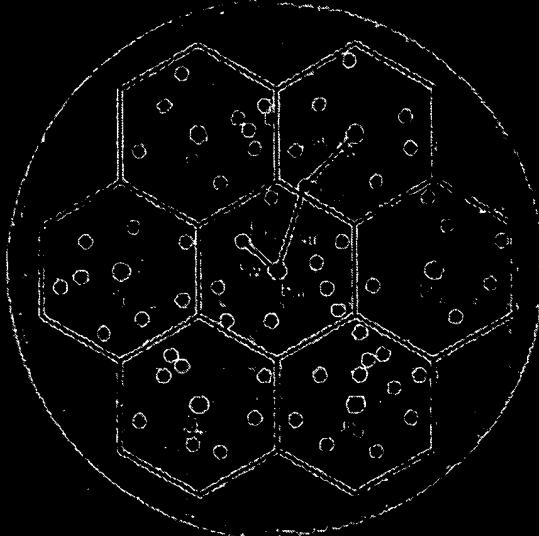
$$\{(60 \times 50 \text{mA-minutes}) / [(60 \text{ minutes/hour} \times 24 \text{ hour/day})(\text{sum of IC currents in mA})]\} \times (24 \text{hour/day})$$

(*) = Unable to locate datasheet for integrated chip (IC) referenced by Schotz

EXHIBIT D

WIRELESS
communications

Principles  Practice



Theodore S. Rappaport

microcellular systems. However, satellite mobile systems offer tremendous promise for paging, data collection, and emergency communications, as well as for global roaming before IMT-2000 is deployed. In early 1990, the aerospace industry demonstrated the first successful launch of a small satellite on a rocket from a jet aircraft. This launch technique is more than an order of magnitude less expensive than conventional ground-based launches and can be deployed quickly, suggesting that a network of LEOs could be rapidly deployed for wireless communications around the globe. Already, several companies have proposed systems and service concepts for worldwide paging, cellular telephone, and emergency navigation and notification [IEE91].

In emerging nations, where existing telephone service is almost nonexistent, fixed cellular telephone systems are being installed at a rapid rate. This is due to the fact that developing nations are finding it is quicker and more affordable to install cellular telephone systems for fixed home use, rather than install wires in neighborhoods which have not yet received telephone connections to the PSTN.

The world is now in the early stages of a major telecommunications revolution that will provide ubiquitous communication access to citizens, wherever they are [Kue91], [Goo91], [ITU94]. This new field requires engineers who can design and develop new wireless systems, make meaningful comparisons of competing systems, and understand the engineering trade-offs that must be made in any system. Such understanding can only be achieved by mastering the fundamental technical concepts of wireless personal communications. These concepts are the subject of the remaining chapters of this text.

1.6 Problems

- 1.1 Why do paging systems need to provide low data rates? How does a low data rate lead to better coverage?
- 1.2 Qualitatively describe how the power-supply requirements differ between mobile and portable cellular phones, as well as the differences between pocket pagers and cordless phones. How does coverage range impact battery life in a mobile radio system?
- 1.3 In simulcasting paging systems, there usually is one dominant signal arriving at the paging receiver. In most, but not all cases, the dominant signal arrives from the transmitter closest to the paging receiver. Explain how the FM capture effect could help reception of the paging receiver. Could the FM capture effect help cellular radio systems? Explain how.
- 1.4 Where would walkie-talkies fit in Tables 1.5 and 1.6? Carefully describe the similarities and differences between walkie-talkies and cordless telephones. Why would consumers expect a much higher grade of service for a cordless telephone system?
- 1.5 Assume a 1 Amp-hour battery is used on a cellular telephone (often called a cellular subscriber unit). Also assume that the phone's radio receiver draws 35 mA on receive and 250 mA during a call. How long would the phone work (i.e., what is the battery life) if the user has one 3-minute call every day? every 6

hours? every hour? What is the maximum talk time available on the cellular phone in this example?

- 1.6 Assume a CT2 subscriber unit has the same size battery as the phone in Problem 1.5, but the paging receiver draws 5 mA and the transmitter draws 80 mA during a call. Recompute the battery life for the cases in Problem 1.5. Recompute the maximum talk time for the CT2 handset.
- 1.7 Why would one expect the CT2 handset in Problem 1.6 to have a smaller battery drain during transmission than a cellular telephone?
- 1.8 Why is F3M, rather than AM, used in most mobile radio systems today? List as many reasons as you can think of, and justify your responses. Consider issues such as fidelity, power consumption, and noise.
- 1.9 List the factors that led to the development of (a) the GSM system for Europe, and (b) the U.S. digital cellular system. How important was it for both efforts to (i) maintain compatibility with existing cellular phones? (ii) obtain spectral efficiency? (iii) obtain new radio spectrum?
- 1.10 Assume that a GSM, an IS-95, and a U.S. digital cellular base station transmit the same power over the same distance. Which system will provide the best SNR at a mobile receiver? How much is the improvement over the other two systems? Assume a perfect receiver with only thermal noise is used for each of the three systems.
- 1.11 Discuss the similarities and difference between a conventional cellular radio system and a space-based cellular radio system. What are the advantages and disadvantages of each system? Which system could support a larger number of users for a given frequency allocation? How would this impact the cost of service for each subscriber?
- 1.12 Assume that wireless communication services can be classified as belonging to one of the following four groups:
 - High power, wide area systems (cellular)
 - Low power, local area systems (cordless telephone and PCS)
 - Low speed, wide area systems (mobile data)
 - High speed, local area systems (wireless LANs)Classify each of the wireless systems described in Chapter 1 using these four groups. Justify your answers. Note that some systems may fit into more than one group.
- 1.13 Discuss the importance of regional and international standards organizations such as ITU-R, ETSI, and WARC. What competitive advantages are there in using different wireless standards in different parts of the world? What disadvantages arise when different standards and different frequencies are used in different parts of the world?
- 1.14 Based on the proliferation of wireless standards throughout the world, discuss how likely it is for IMT-2000 to be adopted. Provide a detailed explanation, along with probable scenarios of services, spectrum allocations, and so on.

Solutions Manual to Accompany

**Wireless Communications
Principles and Practices**

FIRST EDITION

Zhigang Rong

Theodore S. Rappaport



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Cont'd
 infrastructure, complexity, hardware cost are all low.
 A cordless telephone, on the other hand, is a full duplex system. It allows simultaneous two-way communication. Transmission and reception is on two different channels (FDD) although new cordless system are using TDD. The coverage range, required infrastructure, hardware cost of a cordless phone system are high and the complexity is moderate. User expectations are lower for a cordless telephone.

1.5 If the user has one 3-minute call every day
 the battery life = $\frac{60 \times 1000}{(60 \times 6 + 3 \times 15 + 3 \times 50) \text{ (mA-mins)}}$
 $\approx 1.175 \text{ days} \approx 28.2 \text{ hours}$

If the user has one 3-minute call every 6 hours
 the battery life = $\frac{60 \times 1000}{(60 \times 6 + 3 \times 15 + 3 \times 50)} \times 6 \approx 27.8 \text{ hours}$

If the user has one 3-minute call every hour
 the battery life = $\frac{60 \times 1000}{(60 + 3 \times 15 + 3 \times 50)} \approx 21.86 \text{ hours}$
 the maximum talk time = $\frac{60 \times 1000}{250} = 240 \text{ minutes} = 4 \text{ hours}$

1.6 For 3-minute call/day
 battery life = $\frac{60 \times 1000 \text{ (mA-mins)}}{(60 \times 24 + 3 \times 15 + 3 \times 50)} \approx 8.08 \text{ days} = 193.96 \text{ hours}$

1.6 Cont'd

For 3 minute-call / 6 hours

$$\text{battery life} = \frac{60 \times 1000}{(60 \times 6 + 3 \times 15 + 3 \times 50)} \times 6 \approx 177.78 \text{ hours}$$

For 3 minute-call / hour

$$\text{battery life} = \frac{60 \times 1000}{(60 + 3 \times 15 + 3 \times 50)} \approx 114.29 \text{ hours}$$

$$\text{The maximum talk time} = \frac{60 \times 1000}{250} \approx 750 \text{ minutes} = 12.5 \text{ hours}$$

1.7 Since the coverage range of the CT-2 system is lower than that of the cellular radio system, to obtain the same signal-to-noise ratio in the coverage area, a CT-2 handset requires less transmitted power than a cellular telephone, and thus a smaller battery drain.

1.8 FM has several advantages over AM. The most important advantage is FM's superior noise suppression characteristics. With conventional AM, the modulating signal is impressed onto the carrier in the form of amplitude variations. However, noise introduced into the system also produces changes in the amplitude of the envelope. Therefore, the noise cannot be removed from the composite waveform without also removing a portion of the information signal. With FM, the information is impressed onto the carrier in the form of frequency variations. Therefore, with FM receivers,

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